## Hfinturnatk Axsigntent 11



References : all answers are based on information obtained from Wikipedia.

1. 11-1. In physics, what is ether? And what is ethernet?

In physics, ether (or aether) is the fictitious medium that supports light waves. Ethernet:
Ethernet is a family of computer networking technologies used in local area networks (Wikipedia)
Information is transmitted to and from a computer on the network, using a coaxial cable.
2. 11-2. What are the frequencies used by your cell phone?

## Cell phones

My cell phone is an iPhone $8+$ using an AT \& T network. The frequency band (PCS 1900) is

1850-1910 MHz (mobile to base), and 1930-1990 MHz (base to mobile)
3. 11-3. What are the frequencies used for WiFi communication?

## WiFi communication

WLAN (Wireless Local Area Network) channels are frequently accessed using IEEE 802.11 protocols and are sold mostly under the trademark Wi-Fi.
The 802.11 standard provides several distinct radio frequency ranges for use in Wi-Fi communications :
$900 \mathrm{MHz} 2.4 \mathrm{GHz}, 3.6 \mathrm{GHz}, 4.9 \mathrm{GHz}, 5 \mathrm{GHz}, 5.9 \mathrm{GHz}$ and 60 GHz bands.
Channels used in the 2.4 GHz frequency range, in MHz : channel center range

| 4 | 2427 | $2416-2438$ |
| :--- | :--- | :--- |
| 5 | 2432 | $2421-2443$ |
| 6 | 2437 | $2426-2448$ |
| 7 | 2442 | $2431-2453$ |
| 8 | 2447 | $2436-2458$ |
| 9 | 2452 | $2441-2463$ |
| 10 | 2457 | $2446-2468$ |
| 11 | 2462 | $2451-2473$ |

4. 11-4. Jackson Problem 8.2

We did the ideal field calculations for this example in the lecture Monday November 4 .

4 points
5. 11-5. What is the impedance of free space? Explain.

## The impedance of free space

- The impedance of free space $Z_{0}$ is a physical constant relating the magnitudes of the electric and magnetic fields of electromagnetic radiation travelling through free space. That is,

$$
Z_{0}=\frac{|E|}{|H|}
$$

where $|E|$ is the electric field strength and $|H|$ is the magnetic field strength.

- The impedance of free space (for a plane wave in free space)
is equal to the $\mu_{0} c$.
- The presently accepted value is

$$
Z_{0}=376.730313668 \text { (57) ohms. }
$$

A good approximation is $Z_{0} \approx 120 \pi$ ohms.

```
(*verify unit:*)
(* curl H = J = \sigma E; also, R = \rho L/A; *)
(* [H]/[m] = [\sigma][E]; [R]= 1/[\sigma] [m]/[m^2] = [\sigma] -1 [m] -1 ;*)
(* So ... [E]/[H] = [\sigma] [-1 [m] -1 = [R] = ohm *)
(* verify: }\mu0*\textrm{C}=(4\textrm{Pi}\mp@subsup{*}{}{\wedge}-7)*(3*/*^^)=120 Pi *
```

The next problems concern a rectangular waveguide with $\delta \mathrm{x}=\mathrm{a}=5 \mathrm{~cm}$ and $\delta \mathrm{y}=\mathrm{b}$ $=2.5 \mathrm{~cm}$;
also, set $\mu=\mu_{0}$ and $\epsilon=\epsilon_{0}$.
6. 11-6. Calculate the cutoff frequency and the corresponding wavelength for the $\mathrm{TE}_{10}$ mode.

## The $T E_{10}$ mode

The dispersion relation for the wave guide is given by $\gamma^{2}=\mu \epsilon \omega^{2}-k^{2}$.
The $T E_{10}$ mode has $\gamma_{10}=\pi / a$, and the cutoff frequency is $\omega_{10}=\frac{1}{\sqrt{\mu \epsilon}} \gamma_{10}=c \gamma_{10}$.
Thus $\omega_{10}=c \pi / a=1.88 \times 10^{10} \mathrm{~s}^{-1}$.
$3.0 * \wedge 8$ * $\mathrm{Pi} / 0.05$
\% / ( 2 * Pi)
Out[43]= $1.88496 \times 10^{10}$
Out[44]=
3. $\times 10^{9}$

The corresponding wavelength is $\lambda_{10}=\frac{2 \pi c}{\omega_{10}}=2 a=10 \mathrm{~cm}$.

$$
\omega_{10}=1.88 \times 10^{10} / \mathrm{s} \text { or } f_{10}=3.0 \mathrm{GHz} ; \lambda=10 \mathrm{~cm}
$$

7. 11-7. Calculate the energy flux of the $\mathrm{TE}_{10}$ mode.
[Hint: the fields are the Real Parts of the complex functions in (8.46).
Take the real parts before you calculate the Poynting vector. ]

## The energy flux of the $T E_{10}$ mode

The energy flux is $\vec{S}=\vec{E} \times \vec{H}$ where the fields are
$H_{z}=H_{0} \cos (\pi x / a) \cos (k z-\omega t)$
$H_{x}=(k a / \pi) H O \sin (\pi x / a) \sin (k z-\omega t)$
$E_{y}=(-\omega a \mu / \pi) H 0 \sin (\pi x / a) \sin (k z-\omega t)$
$S=\left(\begin{array}{ccc}i & j & k \\ 0 & E y & 0 \\ H x & 0 & H z\end{array}\right)=e x E y H z-e z E y H x$
$S x=H_{0}^{2}(-\omega a \mu / \pi) \sin (\pi x / a) \sin (k z-\omega t) \cos (\pi x / a) \cos (k z-\omega t)$
$S x=(-1 / 4) H_{0}^{2}(\omega a \mu / \pi) \sin (2 \pi x / a) \sin [2(k z-\omega t)]$
$S z=-H_{0}^{2}(-\omega a \mu / \pi) \sin (\pi x / a) \sin (k z-\omega t)(k a / \pi) \sin (\pi x / a) \sin (k z-\omega t)$
$S z=H_{0}^{2}(\omega a \mu / \pi)(k a / \pi) \sin ^{2}(\pi \times / a) \sin ^{2}(k z-\omega t)$
Averaging over the time, $S=\hat{e}_{z}\left\langle S_{z}\right\rangle$ where $S z=H_{0}^{2}\left(\frac{\mu \omega k a^{2}}{2 \pi^{2}}\right) \sin ^{2}\left(\frac{\pi x}{a}\right)$

$$
\left\langle S_{z}\right\rangle=H_{0}^{2}\left(\frac{\mu \omega k a^{2}}{2 \pi^{2}}\right) \sin ^{2}\left(\frac{\pi x}{a}\right)
$$

8. 11-8. Calculate the cutoff frequencies for the TM modes. Hand in a Table like the table below (8.46); the elements of the table should be $\omega_{\text {cutoff }}\left(\mathrm{TM}_{\mathrm{mn}}\right) / \omega_{\text {cutoff }}\left(\mathrm{TE}_{10}\right)$.

## TM modes of the rectangular waveguide

For the TM modes, both $m$ and $n \in\{123 \ldots\}$.
Recall $\gamma^{2}=\mu \epsilon \omega^{2}-k^{2}$. Therefore the cutoff frequencies are
$\omega_{m n}=\gamma_{m n} / \sqrt{\mu \epsilon}=\frac{1}{\sqrt{\mu \epsilon}}\left[(m \pi / a)^{2}+(n \pi / b)^{2}\right]^{1 / 2}$
$\ln [45]:=\operatorname{Do}[D o[t b l[m, n]=$ SetPrecision[
Sqrt $\left.\left[(m * P i)^{\wedge} 2+4 *(n * P i)^{\wedge} 2\right] / P i, 3\right]$,
\{m, 1, 3\}], \{n, 1, 3\}];
row[0] = \{"", "n=1", "n=2", "n=3"\};
col[0] = \{"", "m=1", "m=2", "m=3"\};
Do [
$\operatorname{row}[i]=\{\operatorname{col}[0][[i+1]], \operatorname{tbl}[i, 1], \operatorname{tbl}[i, 2], \operatorname{tbl}[i, 3]\},\{i, 1,3\}] ;$
Style[ TableForm[Join[\{row[0]\}, \{row[1]\}, \{row[2]\}, \{row[3]\}]], \{24, Purple, Bold\}]

|  | $n=1$ | $n=2$ | $n=3$ |
| :--- | :--- | :--- | :--- |
| $m=1$ | 2.24 | 4.12 | 6.08 |
| $m=2$ | 2.83 | 4.47 | 6.32 |
| $m=3$ | 3.61 | 5.00 | 6.7 |

9. 11-9. For the waveguide mode $\mathrm{TE}_{32} \ldots$
hand in a sketch (better: a computer graphic) of the effective surface current density $\boldsymbol{K}(\mathrm{x}, \mathrm{y})$ at the wall of the waveguide with $y=0$, for $\omega=2 \omega_{32}$.

## The $T E_{32}$ waveguide mode

The goal is to calculate the surface current density $K(x, y)$ at $y=0$.
Recall the boundary condition: $\Delta H_{\text {tangential }}=\mathrm{K} \times \hat{n}$.
For a perfect conductor, $\mathrm{H}=\mathrm{O}$ inside the conductor; so the boundary condition is $H_{\text {tangential }}=K \times \hat{n}$. For the wall at $y=0$, the normal direction is $-\hat{e}_{y}$, and the tangential directions are $\hat{\boldsymbol{e}}_{x}$ and $\hat{\boldsymbol{e}}_{z}$.
To calculate: $K=\hat{n} \times\left(\hat{\boldsymbol{e}}_{x} H_{x}+\hat{\boldsymbol{e}}_{z} H_{z}\right)=\hat{\boldsymbol{e}}_{z} H_{x}-\hat{\boldsymbol{e}}_{x} H_{z}$ (evaluated at $y=0$ )
Fields of the $T E_{32}$ mode:

$$
\begin{aligned}
& H_{z}=H_{0} \cos \left(\frac{3 \pi x}{a}\right) \cos \left(\frac{2 \pi y}{a}\right) \cos (k z-\omega t)=H_{0} \cos \left(\frac{3 \pi x}{a}\right) \cos (k z-\omega t) @_{y}=0 \\
& H_{1}=\frac{k H_{0}}{} \frac{3 \pi}{\sin \left(\frac{3 \pi x}{}\right)} \cos \left(\frac{2 \pi y}{}\right) \sin \left(k 7-(1, t)=\frac{k H_{0}}{} \frac{3 \pi}{\sin }\left(\left.\frac{3 \pi x}{}\right|_{\sin (k 7-(1, t)} @_{v}=0\right.\right.
\end{aligned}
$$

$$
H_{y}=\frac{k H_{0}}{y^{2}} \frac{2 \pi}{b} \cos \left(\frac{3 \pi x}{a}\right) \sin \left(\frac{2 \pi y}{a}\right) \sin (k z-\omega t)=0 @ y=0
$$

The current density on the wall at $\mathrm{y}=0$
$K_{x}=-H_{z}=H_{0} \cos \left(\frac{3 \pi x}{a}\right) \cos (k z-\omega t)$
$K_{z}=H_{x}=H_{0} \frac{3 \pi k}{a \gamma^{2}} \sin \left(\frac{3 \pi x}{a}\right) \sin (k z-\omega t)$;
here, $\gamma^{2}=(3 \pi / a)^{2}+(2 \pi / b)^{2}$ and $\gamma^{2}=\mu \epsilon \omega^{2}-k^{2}$.
Numerical parameters
$a=5 \mathrm{~cm} ; b=2.5 \mathrm{~cm} ; \mu=\mu_{0} ; \epsilon=\epsilon_{0}$
$\gamma=\pi \mathrm{cm}^{-1} ; \omega=2 \mathrm{r} / \mathrm{c}=2 \pi / \mathrm{ccm}^{-1} ; \mathrm{k}=\sqrt{3} \pi \mathrm{~cm}^{-1}$
$K_{x}=H_{0} \cos [3 \pi x /(5 \mathrm{~cm})] \cos (k z-\omega t)$
$K_{z}=H_{0}\left(\frac{3 \sqrt{3}}{5}\right) \sin [3 \pi x /(5 \mathrm{~cm})] \sin (k z-\omega t)$
$\ln [\rho]:=\left\{\operatorname{Sqrt}\left[(3 * \operatorname{Pi} / 5)^{\wedge} 2+(2 * \operatorname{Pi} / 2.5)^{\wedge} 2\right], 3^{\wedge} 2+4^{\wedge} 2, \operatorname{Sqrt}\left[4 * \gamma^{\wedge} 2-\gamma^{\wedge} 2\right] / \cdot\{\gamma \rightarrow \operatorname{Pi}\}\right\}$
Out $[-]=\{3.14159,25, \sqrt{3} \pi\}$
$\ln [\rho]:=\mathrm{Kx}\left[\mathrm{x}_{-}, \mathrm{z}_{-}\right]=\operatorname{Cos}[3 * \mathrm{Pi} * \mathrm{x} / 5]$ * $\operatorname{Cos}[\operatorname{Sqrt}[3] * \operatorname{Pi} * \mathrm{z}] ;$
$\mathrm{Kz}\left[\mathrm{x}_{-}, \mathrm{z}_{-}\right]=(3$ * Sqrt[3]/5) *Sin[3*Pi*x/5] *Sin[Sqrt[3] *Pi*z];
StreamPlot $[\{K x[x, z], K z[x, z]\},\{x, 0,5\},\{z, 0,5\}$,
PlotRange $\rightarrow\{\{0,5\},\{0,5\}\}$, AspectRatio $\rightarrow 1$, StreamPoints $\rightarrow$ Automatic,
Frame $\rightarrow$ True, FrameLabel $\rightarrow$ \{"x", "z"\}, BaseStyle $\rightarrow$ 18]

10. 11-10. Consider the $T E_{m n}$ wave with $\omega>\omega_{m n}$.
(a) Calculate the phase velocity and show that it is is greater than c .
(b) Calculate the group velocity and show that $v_{\text {group }}=c^{2} / v_{\text {phase }}$.

## Phase velocity and group velocity

Assume $\mu=\mu_{0}$ and $\epsilon=\epsilon_{0}$.
Recall $\gamma^{2}=\mu_{0} \epsilon_{0} \omega^{2}-k^{2}=\mu_{0} \epsilon_{0} \omega_{m n}^{2}$.
Therefore, $\omega^{2}=\omega_{m n}^{2}+c^{2} k^{2}$.
(A) The phase velocity is $v_{\text {phase }}=\frac{\omega}{k}=\frac{c \omega}{\sqrt{\omega^{2}-\omega_{m n}^{2}}}>C$
(B) The group velocity is $v_{\text {group }}=\frac{\mathrm{d} \omega}{\mathrm{dk}}=\frac{c^{2} k}{\omega}=\frac{c \sqrt{\omega^{2}-\omega_{m n}^{2}}}{\omega}=\frac{c^{2}}{v_{\text {phase }}}$.
(A) $\omega / k$
AND
(B) $c^{2} k / \omega$

4 points
11. 11-11. Why is a waveguide sometimes better than a coaxial cable? Explain, and define "better".

Both waveguides and coaxial cables can be used to carry energy and information in microwaves.
When is a waveguide better than a coaxial cable? From a Google search:

- The waveguide has no inner conductor, so it is easier to manufacture.
- The waveguide has less energy loss than a coaxial line:
- no power is lost through radiation
- no dielectric loss
- The waveguide can handle higher power than a coaxial cable:

The interior of the waveguide is air, and the breakdown voltage is $30 \mathrm{kV} / \mathrm{cm}$; whereas there is a dielectric material in the coaxial cable with a lower breakdown voltage.
These are advantages of waveguide but there are also disadvantages, so sometimes the coaxial cable is better.

